

Homework No. 05

Lecture discussed the use of volume and surface equivalent on the analysis of cylindrical monopole antenna over a ground plane that is fed by a connected coaxial line. This homework recapitulates the description and addresses the solution of input impedance via field integral equation.

Setup: In order to simplify the analysis, let us assume that only the dominant TEM mode exists in the coaxial aperture. That is, the electric field in the coaxial aperture reads

$$\mathbf{E} = \rho_0 \frac{U}{\rho \ln \frac{b}{a}}, \quad (1)$$

where a, b is the inner and outer radius of the coaxial line, U is the voltage at the aperture and where coaxial line runs along z -axis. The monopole is assumed to be a continuation of the inner coaxial conductor ($z \in (0, L/2)$) which has a form of a thin-wall perfectly electrically conducting (PEC) tube. No cap is assumed at the end of the monopole. The ground plane is assumed to be perfectly conducting and positioned at $z = 0$ plane.

In order to obtain the integral equation for the current on the monopole, a surface equivalent is applied on the half space $z < 0$ with a PEC filling while volume equivalence (or surface equivalence with vacuum filling) is applied on the monopole. Applying further image theory over the PEC surface at $z = 0$ we are finally left with an electric current

$$\mathbf{J}_e = I(z) \frac{\delta(\rho - a)}{2\pi\rho} \mathbf{z}_0, \quad z \in (-L/2, L/2), \quad (2)$$

which is at xy plane encircled by a magnetic current

$$\mathbf{J}_m = -\frac{2U}{\rho \ln \frac{b}{a}} \delta(z) \boldsymbol{\varphi}_0, \quad \rho \in (a, b). \quad (3)$$

Notice that the current $I(z)$ on the monopole (due to reflection it is actually a dipole) must be an even function due to the applied reflection through the PEC plane.

The current $I(z)$ is unknown and is to be obtained from boundary condition on electric field over the surface of the monopole (dipole) tube. Particularly, there must be

$$\boldsymbol{\rho}_0 \times [\mathbf{E}_e(\rho = a, z) + \mathbf{E}_m(\rho = a, z)] = 0, \quad z \in (-L/2, L/2), \quad (4)$$

where it is important to realize that current density $\mathbf{J}_e$ and $\mathbf{J}_m$ radiate in free space and, therefore, the electric field $\mathbf{E}_e, \mathbf{E}_m$ produced by them is relatively easy to find.

Task No. 1: Study [1, 4.8] and show that

$$\mathbf{z}_0 \cdot \mathbf{E}_m = \frac{jU}{\ln \frac{b}{a}} \int_0^\infty \frac{e^{-jk_z|z|}}{k_z} \left[J_0(k_\rho b) - J_0(k_\rho a) \right] J_0(k_\rho \rho) k_\rho dk_\rho, \quad (5)$$

where $k_z^2 = k^2 - k_\rho^2$, $\text{Im}\{k_z\} < 0$, k denotes free-space wavenumber and J_n denotes Bessel's function of order n .

Task No. 2: Use Fourier's transform along z on Helmholtz's equation for magnetic vector potential generated by current $\mathbf{J}_e$ together with [1, 4.7] and show that

$$\mathbf{z}_0 \cdot \mathbf{E}_e(\rho > a, z) = -\frac{kZ}{8\pi} \int_{-\infty}^\infty \left(1 - \frac{k_z^2}{k^2} \right) \tilde{I}(k_z) H_0^{(2)}(k_\rho \rho) J_0(k_\rho a) e^{jk_z z} dk_z, \quad (6)$$

where Z is the free-space impedance, $\tilde{I}(k_z) = \mathcal{F}_z\{I(z)\}$ is the Fourier's transform of the current on the monopole (dipole), $H_n^{(2)}$ is Hankel's function of second kind and order n and $k_\rho^2 = k^2 - k_z^2$ with $\text{Im}\{k_\rho\} < 0$.

Task No. 3: Use (4) to formulate integral equation for unknown current $I(z)$ and then apply method of moments to solve this equation using expansion

$$I(z) = \sum_{n=1,3,5,\dots} I_n \sin \left(n\pi \left(\frac{z}{L} - \frac{1}{2} \right) \right). \quad (7)$$

Evaluate input impedance of the monopole as seen from the coaxial line as

$$Z_{\text{in}} = \frac{U}{I(0)}. \quad (8)$$

For numerical evaluation use $kL \approx 0.9\pi$, $L/a = 50$ and $b/a = 2.3$. You should observe that $Z_{\text{in}} \approx 37\Omega$. Plot the current $I(z)$ along the dipole under these conditions.

References

- [1] D. G. Dudley, *Mathematical Foundations for Electromagnetic Theory*. IEEE Press, 1994.